

suggestion that the short-period hard gamma-ray intensity increases or decreases with the delayed neutrons in going from one fissionable isotope to another. In order to determine the exact nature of the association between

the delayed neutrons and the short-period hard gamma rays, it would be helpful to examine the gamma radiations from quick chemical separations of individual delayed neutron emitters.

## Dielectric Properties of a Lattice of Anisotropic Particles\*

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The constitutive dielectric parameters for uniform space arrays of generalized structural geometry composed of similarly oriented elements of completely generalized material and shape are derived. The theoretical procedure employed in evaluating these parameters is analogous to the classical method applied to the study of the dielectric properties of nonpolar media and assumes that the disturbing action of each particle on a uniform static field can be allowed for, if each particle is replaced by a set of dipoles. Consequently, the results are applicable to microwave frequencies only if the cross-sectional dimensions and interelemental spacings remain small compared to the wavelength. The general result is particularized for the case in which the elements are spherical objects with a tensor permeability and scalar dielectric constant. The results have several applications to the design of artificial dielectrics and some of these are given.

### INTRODUCTION

THE dielectric properties of lattices composed of identical metallic or dielectric elements of various geometries, such as spheres, disks, and strips have been investigated from a molecular point of view by Kock,<sup>1</sup> Corkum,<sup>2</sup> and others. These investigations have treated two cases in both of which the element size and spacings are small compared to the wavelength. The first applies to the case where the spacings are large compared to element size and which therefore neglects interaction effects. The second treats interaction for the special case in which the lattice has the structural isotropy of a cubical array and for which the application of the Clausius-Mosotti relation is valid, when the elements are not too closely packed.

The main objective of this paper will be to extend the treatment to general uniform lattice structures made of identically shaped and oriented particles of general constitutive characteristics. Thus, it will include the

most general case of a uniform lattice with structural anisotropy and both element isotropy and anisotropy at the lattice points.

It is known that if a uniform electric field of strength  $\mathbf{E}_0$  is applied to an array of like elements (see Fig. 1), a distribution of charge is established within each element. The total charge of each element remains zero but is redistributed in such a way that an additional field is created. A first approximation to this extra field is obtained by examining the charge distribution of an isolated element in a uniform field. This induced field can, for a generalized element, be considered as being approximately equivalent to a set of three dipoles respectively parallel to each of the coordinate axes. The resultant of these dipoles, the electric dipole moment vector, will be denoted by  $\mathbf{p} = p_x \mathbf{a}_x + p_y \mathbf{a}_y + p_z \mathbf{a}_z$  where  $\mathbf{a}_x$ ,  $\mathbf{a}_y$ , and  $\mathbf{a}_z$  are the unit coordinate vectors. The magnitude of  $\mathbf{p}$  is proportional to the magnitude of the applied field  $\mathbf{E}_0$ ,

$$\mathbf{p} = (\alpha) \cdot \mathbf{E}_0. \quad (1)$$

The constant of proportionality ( $\alpha$ ) is the polarizability tensor and its value depends on the geometry and material of the element. The polarization  $\mathbf{P}$  is the total dipole moment per unit volume, so that if there are  $N$  elements per unit volume

$$\mathbf{P} = N\mathbf{p} = N(\alpha) \cdot \mathbf{E}_0. \quad (2)$$

The dielectric constant tensor ( $k_e$ ), is related to the electric field strength and the polarization by the equation

$$\mathbf{D} = (\epsilon) \cdot \mathbf{E}_0 = (\epsilon_0) \cdot \mathbf{E}_0 + \mathbf{P}, \quad (3)$$

where  $\mathbf{D}$  = the displacement vector,  $\epsilon_0$  = permittivity of free space,  $\epsilon$  = permittivity of the lattice medium

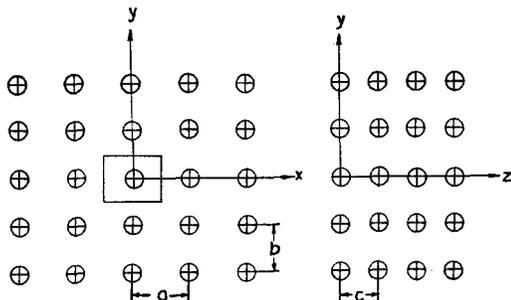


FIG. 1. A tetragonal array.

\* This work was supported by the Office of Naval Research.

<sup>1</sup> W. E. Kock, Bell System Tech. J. 27, 58 (1948).

<sup>2</sup> R. W. Corkum, Proc. Inst. Radio Engrs. 40, 574 (1952).

( $\epsilon = \epsilon_0 k_e$ ), and  $k_e$  = dielectric constant of the lattice medium. Substitution for  $\mathbf{p}$  from Eq. (2) yields

$$(k_e) = (1) + [N(\alpha)/\epsilon_0]. \quad (4)$$

In the foregoing, it was assumed that the field acting on each individual element in the presence of the others remained equal to the externally applied field and that the charge distribution of each element corresponds to that of a set of dipoles. The first restriction will be removed in the next section by including the contributions of all elements in the exciting field of each element. The second restriction implies that the elements of the array are not too closely packed, and this will be assumed to be the case in the following. That is, the packing is so close that interactions must be considered but not so close that the resulting field distortion can no longer be described by simple dipole fields.

### I. Evaluation of the Dielectric Constant of an Array of Isotropic Elements

The method used to take interaction into consideration is a standard one used in the molecular theory of solids and consists simply of calculating the additional field acting on each element because of the dipoles induced in each of the other elements. To consider the interaction field  $\mathbf{E}_s$ , consider a dipole with charges  $\pm q$  separated by a distance  $d$  (see Fig. 2). The potential  $\varphi$  at a distance  $r$  from its center is given by

$$\varphi = \frac{q}{4\pi\epsilon_0} \left( \frac{1}{r_+} - \frac{1}{r_-} \right).$$

For  $r \gg d$ ,  $r_- - r_+ \simeq d \cos\theta$ ,  $r_+ r_- = r^2$ , and hence

$$\varphi = \frac{qd}{4\pi\epsilon_0 r^2} \cos\theta = \frac{\mathbf{p} \cdot \mathbf{r}}{4\pi\epsilon_0 r^3} \quad (qd = \mathbf{p}). \quad (5)$$

The field at a distance  $\mathbf{r}$  from this dipole is

$$\begin{aligned} \mathbf{E} &= -\frac{1}{4\pi\epsilon_0} \nabla \left( \frac{\mathbf{p} \cdot \mathbf{r}}{r^3} \right) \\ &= -\frac{1}{4\pi\epsilon_0} \left[ \frac{\mathbf{p}}{r^3} - \frac{3(\mathbf{p} \cdot \mathbf{r})\mathbf{r}}{r^5} \right]. \end{aligned} \quad (6)$$

This can be expanded in rectangular coordinates as

$$\begin{aligned} E_x &= \frac{1}{4\pi\epsilon_0} \left( \frac{3x^2 - r^2}{r^5} p_x + \frac{3xy}{r^5} p_y + \frac{3xz}{r^5} p_z \right) \\ E_y &= \frac{1}{4\pi\epsilon_0} \left( \frac{3xy}{r^5} p_x + \frac{3y^2 - r^2}{r^5} p_y + \frac{3yz}{r^5} p_z \right) \\ E_z &= \frac{1}{4\pi\epsilon_0} \left( \frac{3xz}{r^5} p_x + \frac{3yz}{r^5} p_y + \frac{3z^2 - r^2}{r^5} p_z \right). \end{aligned}$$

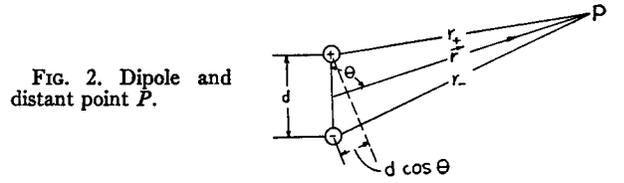


FIG. 2. Dipole and distant point  $P$ .

It follows that each component of  $\mathbf{E}$  is a linear function of the components of  $\mathbf{p}$ , the coefficients being functions of the components  $x, y, z$  of the radius vector  $\mathbf{r} = (x^2 + y^2 + z^2)^{1/2}$ . The interaction field  $\mathbf{E}_s$ , resulting from an infinite array of elements can be computed by considering the element at the origin of coordinates removed (see Fig. 1), the array being otherwise unchanged, and the contributions of the infinite number of dipole elements summed up at the origin. The interaction field  $\mathbf{E}_s$  is therefore given by

$$\mathbf{E}_s = (T) \cdot \mathbf{p}, \quad (7)$$

where  $(T)$  is a symmetric tensor whose matrix is

$$(T) = \frac{1}{4\pi\epsilon_0} \begin{bmatrix} \sum_i \frac{3x^2 - r^2}{r^5} & \sum_i \frac{3xy}{r^5} & \sum_i \frac{3xz}{r^5} \\ \sum_i \frac{3xy}{r^5} & \sum_i \frac{3y^2 - r^2}{r^5} & \sum_i \frac{3yz}{r^5} \\ \sum_i \frac{3xz}{r^5} & \sum_i \frac{3yz}{r^5} & \sum_i \frac{3z^2 - r^2}{r^5} \end{bmatrix},$$

$i$  is the summation index, and the sums are extended over all elements within the array except the single element at the center of coordinates. For the case when the elements themselves are isotropic the induced dipole moment vector will be given by

$$\mathbf{p} = \alpha(\mathbf{E}_0 + \mathbf{E}_s) = \alpha[(T) \cdot \mathbf{p} + \mathbf{E}_0], \quad (8)$$

where  $\alpha$  as mentioned earlier is the electric polarizability of the isotropic elements and is a scalar quantity. If Eq. (8) is solved for  $\mathbf{p}$ , there results:

$$\mathbf{p} = - \left[ (T) - \left( \frac{1}{\alpha} \right) \right]^{-1} \cdot \mathbf{E}_0 \quad (9)$$

where  $(1/\alpha)$  is the matrix  $1/\alpha$ . Let  $R$  denote the matrix  $(T) - (1/\alpha)$  then,

$$\mathbf{p} = (R^{-1}) \cdot \mathbf{E}_0, \quad (10)$$

where

$$(R) = \begin{bmatrix} T_{xx} - \frac{1}{\alpha} & T_{xy} & T_{xz} \\ T_{yx} & T_{yy} - \frac{1}{\alpha} & T_{yz} \\ T_{zx} & T_{zy} & T_{zz} - \frac{1}{\alpha} \end{bmatrix}.$$

Substitution of the value of  $\mathbf{p}$  in the following equation

$$(\epsilon) \cdot \mathbf{E}_0 = \epsilon_0 \mathbf{E}_0 + N\mathbf{p} \quad (11)$$

yields

$$(\mathbf{k}_e) - (1) = -\frac{N}{\epsilon_0} (R^{-1}), \quad (12)$$

where (1) is the unitary matrix. The inverse of the matrix (R) is given by

$$(R^{-1}) = \frac{(R^{ij})}{\det(R)}. \quad (13)$$

The matrix  $(R^{ij})$  is formed by arranging the cofactors of the elements  $R_{ij}$  in a matrix array and then transposing the rows and columns of the resulting matrix. Since the elements were assumed to be isotropic, it is important to observe that the tensor nature of the dielectric constant is entirely due to the geometry of the lattice medium.

#### Tetragonal Lattice

For a tetragonal array, it can easily be shown that the elements  $T_{xy}, T_{xz}, T_{yz}, T_{yx}, T_{zx}, T_{zy}$  are zero, and the dielectric constant reduces to a diagonal tensor of the form

$$(\mathbf{k}_e) = (1) - \frac{N}{\epsilon_0} \begin{bmatrix} \frac{1}{\alpha} & 0 & 0 \\ T_{xx} & \frac{1}{\alpha} & 0 \\ 0 & T_{yy} & \frac{1}{\alpha} \\ 0 & 0 & T_{zz} \end{bmatrix}; \quad (14)$$

and therefore

$$\begin{aligned} k_{exx} &= 1 + \frac{N\alpha/\epsilon_0}{1 - \alpha T_{xx}}, & k_{eyy} &= 1 + \frac{N\alpha/\epsilon_0}{1 - \alpha T_{yy}}, \\ k_{ezz} &= 1 + \frac{N\alpha/\epsilon_0}{1 - \alpha T_{zz}}. \end{aligned} \quad (15)$$

In which, for the tetragonal array,

$$T_{xx} = \frac{1}{4\pi\epsilon_0} \sum'_{m_1=-\infty} \sum'_{m_2=-\infty} \sum'_{m_3=-\infty} \frac{2m_1a^2 - b^2m_2^2 - c^2m_3^2}{(m_1^2a^2 + m_2^2b^2 + m_3^2c^2)^{5/2}}, \quad (16)$$

where  $a, b,$  and  $c$  are the interelemental spacings as shown in Fig. 1 and  $m_1, m_2,$  and  $m_3$  are any integers. The prime signifies the omission of the term for which  $m_1, m_2,$  and  $m_3$  are all zero. Unfortunately, this series is only conditionally convergent, and different values can be obtained for its sum depending on the order in

which the summations are performed. Lorentz<sup>3</sup> has used a convenient scheme by which he replaces the collection of dipoles by a continuous dielectric and then removing a sphere at the origin. This series will be considered again later in the discussion and at present the special case of a cubical lattice where  $a=b=c$  will be considered in detail.

#### Cubical Lattice

For a cubical lattice  $T_{xx}$  becomes:

$$T_{xx} = \frac{1}{a^3 4\pi\epsilon_0} \sum'_{m_1=-\infty} \sum'_{m_2=-\infty} \sum'_{m_3=-\infty} \frac{2m_1^2 - m_2^2 - m_3^2}{(m_1^2 + m_2^2 + m_3^2)^{5/2}},$$

where  $a$  is the interelemental spacing. This can be written as

$$\begin{aligned} T_{xx} &= \frac{1}{4\pi\epsilon_0 a^3} \left\{ 2 \sum_{m_1=1}^{\infty} \frac{2}{m_1^3} - 2 \sum_{m_2=1}^{\infty} \frac{1}{m_2^3} - 2 \sum_{m_3=1}^{\infty} \frac{1}{m_3^3} \right\} \\ &+ \frac{2}{\pi\epsilon_0 a^3} \sum_{m_1=1}^{\infty} \sum_{m_2=1}^{\infty} \sum_{m_3=1}^{\infty} \frac{2m_1^2 - m_2^2 - m_3^2}{(m_1^2 + m_2^2 + m_3^2)^{5/2}}. \end{aligned} \quad (17)$$

The first three single index series cancel each other and  $T_{xx}$  becomes

$$T_{xx} = \frac{2}{\pi\epsilon_0 a^3} \sum_{m_1=1}^{\infty} \sum_{m_2=1}^{\infty} \sum_{m_3=1}^{\infty} \frac{2m_1^2 - m_2^2 - m_3^2}{(m_1^2 + m_2^2 + m_3^2)^{5/2}}. \quad (18)$$

To evaluate the above series, the Poisson transformation is used. For a triple series this transformation can be written as

$$\begin{aligned} \sum_a^{\infty} \sum_b^{\infty} \sum_c^{\infty} f(m_1, m_2, m_3) &= -\frac{f(a, b, c)}{8} \\ &+ \frac{1}{2} \left\{ \sum_{m_2=b}^{\infty} \sum_{m_3=c}^{\infty} f(a, m_2, m_3) \right. \\ &+ \sum_{m_1=a}^{\infty} \sum_{m_3=c}^{\infty} f(m_1, b, m_3) + \sum_{m_1=a}^{\infty} \sum_{m_2=b}^{\infty} f(m_1, m_2, c) \left. \right\} \\ &+ \int_a^{\infty} \int_b^{\infty} \int_c^{\infty} f(\alpha, \beta, \gamma) d\alpha d\beta d\gamma \\ &+ 8 \sum_{n_1=1}^{\infty} \sum_{n_2=1}^{\infty} \sum_{n_3=1}^{\infty} \int_a^{\infty} \int_b^{\infty} \int_c^{\infty} f(\alpha, \beta, \gamma) \\ &\quad \times \cos 2\pi n_1 \alpha \cos 2\pi n_2 \beta \cos 2\pi n_3 \gamma d\alpha d\beta d\gamma. \end{aligned} \quad (19)$$

The symmetry of the expression causes the first term and the double index series to cancel. The triple summation of the Poisson transformation formula

<sup>3</sup>H. A. Lorentz, *Theory of Electrons* (B. G. Teubner, Leipzig, Germany, 1909).

represents corrections to the trapezoidal rule and is negligible in this case, because the expression for  $T_{zz}$  varies slowly with respect to  $m_1$ ,  $m_2$ , and  $m_3$ . Hence,

$$T_{zz} = \frac{2}{\pi \epsilon_0 a^3} \int_{\frac{1}{2}}^{\infty} \int_{\frac{1}{2}}^{\infty} \int_{\frac{1}{2}}^{\infty} \frac{2\alpha^2 - \beta^2 - \gamma^2}{(\alpha^2 + \beta^2 + \gamma^2)^{5/2}} d\alpha d\beta d\gamma. \quad (20)$$

Carrying on the integration with respect to  $\alpha$ ,

$$\begin{aligned} T_{zz} &= \frac{2}{\pi \epsilon_0 a^3} \int_{\frac{1}{2}}^{\infty} \int_{\frac{1}{2}}^{\infty} \left[ \frac{-\alpha}{(\alpha^2 + \beta^2 + \gamma^2)^{\frac{3}{2}}} \right]_{\frac{1}{2}}^{\infty} d\beta d\gamma \\ &= \frac{1}{\pi \epsilon_0 a^3} \int_{\frac{1}{2}}^{\infty} \int_{\frac{1}{2}}^{\infty} \frac{d\beta d\gamma}{(\frac{1}{4} + \beta^2 + \gamma^2)^{\frac{3}{2}}}. \end{aligned} \quad (21)$$

Using the formula,

$$\int \frac{dx}{(a+bx^2)^{m+1}} = \frac{1}{2ma} \frac{x}{(a+bx^2)^m} + \frac{2m-1}{2ma^2} \int \frac{dx}{(a+bx^2)^m},$$

Eq. (21) becomes

$$\begin{aligned} T_{zz} &= \frac{1}{\pi \epsilon_0 a^3} \int_{\frac{1}{2}}^{\infty} \left[ \frac{1}{(\frac{1}{4} + \gamma^2)} \frac{\beta}{(1 + \beta^2 + \gamma^2)^{\frac{3}{2}}} \right]_{\frac{1}{2}}^{\infty} d\gamma \\ &= \frac{1}{\pi \epsilon_0 a^3} \int_{\frac{1}{2}}^{\infty} \left[ \frac{1}{\frac{1}{4} + \gamma^2} \frac{1/2}{(\frac{1}{4} + \gamma^2)(\frac{1}{2} + \gamma^2)^{\frac{3}{2}}} \right] d\gamma. \end{aligned} \quad (22)$$

Making the substitution

$$\frac{1}{4} + \gamma^2 = u^2$$

in the second term of the integral, there results

$$\begin{aligned} T_{zz} &= \frac{1}{\pi \epsilon_0 a^3} \left[ \int_{\frac{1}{2}}^{\infty} \frac{1}{\frac{1}{4} + \gamma^2} d\gamma - \frac{1}{2} \int_{\frac{1}{2}}^{\infty} \frac{du}{u(u^2 - \frac{1}{16})^{\frac{3}{2}}} \right] \\ &= \frac{1}{\pi \epsilon_0 a^3} \left\{ 2 \tan^{-1} \frac{\gamma}{1/2} \Big|_{\frac{1}{2}}^{\infty} - \cos^{-1} \frac{1/4}{u^2} \Big|_{1/\sqrt{2}}^{\infty} \right\} \\ &= \frac{1}{\pi \epsilon_0 a^3} \frac{\pi}{3} = \frac{N}{3 \epsilon_0} \quad \left( \text{since } N = \frac{1}{a^3} \right). \end{aligned} \quad (23)$$

Substitution of Eq. (23) in Eq. (15) yields:

$$k_{ezz} = 1 + \frac{N\alpha/\epsilon_0}{1 - (\alpha N/3\epsilon_0)}.$$

The expression for  $k_{ezz}$  is often written

$$\frac{k_{ezz} - 1}{k_{ezz} + 2} = \frac{N\alpha}{3\epsilon_0} \quad (24)$$

and is known as the Clausius-Mosotti formula. This formula has been used to determine the microwave properties of artificial dielectrics when the dimensions and spacings of the elements are small with respect to wavelength and the geometry of the obstacles possesses a three-dimensional symmetry which permits the evaluation of averaged constitutive parameters. It is

important to realize that Eq. (24) is valid only when the lattice is cubical, otherwise the application of the Clausius-Mosotti relation will be an approximation.<sup>4</sup>

#### Evaluation of the Tensor Components

The usefulness of Eq. (12) for the determination of the dielectric constants of lattices of general shape depends on the tractability of the summations expressing the tensor components. However, for any infinite uniform lattice, it is possible to approximate the summation by the following artifice. Consider the lattice to be a *cubical* one except in the neighborhood of the origin of coordinates at which point the interaction field  $\mathbf{E}_s$  is being computed. The physical justification for this is that, for a central region sufficiently large, the value of the summation of the elements outside does not appreciably depend on small deviations from a cubical pattern. Furthermore, when the array is cubical, the contribution of the interaction field  $\mathbf{E}_s$  is zero within a *finite* sphere or cube around the origin. Therefore if we excise the central noncubical region and substitute for it a cubical one, we have added no contribution to the remainder of the array. As a consequence of this reasoning, it is clear that any uniform lattice can be replaced by an infinite cubical lattice with interelemental spacing  $(1/\sqrt[3]{N})$ , plus a finite spherical or cubical region with the original lattice pattern and centered around the point where the interaction field is being evaluated (see Fig. 3). Therefore the summations expressed by the components of the structural anisotropy tensor ( $T$ ) will consist of two parts, (a) the contri-

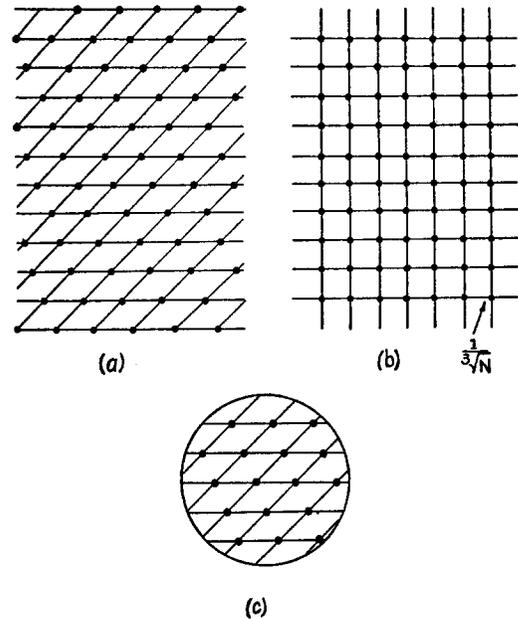


FIG. 3. (a) An infinite uniform lattice = (b) an infinite cubical lattice + (c) a finite region of the lattice shown in (a).

<sup>4</sup> J. Brown and W. Jackson, Proc. Inst. Elec. Engrs. (London) 107, 37 (1955).

bution of the infinite cubical array to be denoted by  $T_{ij}'$  and (b) the contribution of the finite region with the original lattice pattern, to be denoted by  $T_{ij}''$ . The values of primed quantities can be expressed by

$$T_{ij}' = \frac{N}{3\epsilon_0} \delta(i-j) \left[ \text{see Eqs. (16)-(23)} \right] \left. \begin{matrix} i \\ j \end{matrix} \right\} = x, y, z,$$

where  $\delta(i-j)$  is the delta function. To evaluate  $T_{ij}''$  the summation must be carried on term by term within the finite sphere or cube centered around the origin. It should be noted that this finite region can be surprisingly small and still render the approximation valid, since the greatest contribution to the interaction field find their source in elements very near the origin.<sup>5</sup> For a tetragonal array  $a=b=5$  units and  $c=1$  unit, the values of  $T_{zz}''$  for different trial radii are:

$$\begin{aligned} 4\pi\epsilon_0 T_{zz}'' &= -1.182152, & r &= (1200)^{\frac{1}{3}} \text{ units} \\ 4\pi\epsilon_0 T_{zz}'' &= -1.181886, & r &= (1875)^{\frac{1}{3}} \text{ units.} \end{aligned}$$

For a cube

$$4\pi\epsilon_0 T_{zz}'' = -1.181838 \quad \text{side 30 units.}$$

It should also be noted that

$$T_{xx}'' + T_{yy}'' + T_{zz}'' = \frac{1}{4\pi\epsilon_0} \sum_i \frac{3x^2 - r^2}{r^5} + \frac{3y^2 - r^2}{r^5} + \frac{3z^2 - r^2}{r^5} = 0. \quad (25)$$

An application of this method will be illustrated in Sec. II, Example 2.

## II. Evaluation of the Dielectric Constant of an Array of Anisotropic Elements

In the previous section, it was shown that for a general lattice geometry, the dielectric constant is a tensor which reduces to a diagonal tensor for a tetragonal array and to a scalar in the case of a cubical arrangement. The expressions derived are valid only for the case where the lattice elements themselves are isotropic. When the assumption of element isotropy is removed, however,  $\alpha$ , (the polarizability of these elements) becomes itself a tensor. The expression for the dielectric tensor components derived in the previous section will now be generalized to include the anisotropy of the elements.

When  $\alpha$  is a tensor, Eq. (8) becomes

$$\mathbf{p} = (\alpha) \cdot [(T) \cdot \mathbf{p} + \mathbf{E}_0], \quad (26)$$

where

$$(\alpha) = \begin{bmatrix} \alpha_{xx} & \alpha_{xy} & \alpha_{xz} \\ \alpha_{yx} & \alpha_{yy} & \alpha_{yz} \\ \alpha_{zx} & \alpha_{zy} & \alpha_{zz} \end{bmatrix}.$$

Solving for  $\mathbf{p}$ , yields

$$\mathbf{p} = -[(\alpha)(T) - (1)]^{-1}(\alpha) \cdot \mathbf{E}_0.$$

Using the abbreviations  $(\alpha)(T) - (1) = (S)$ , and sub-

stitution of the value of  $\mathbf{p}$  in Eq. (11) yields

$$(k_e) - (1) = -\frac{N}{\epsilon_0} (S^{-1})(\alpha). \quad (27)$$

### Example 1. Cubical Array of Ferrite Spheres

Element anisotropy can be realized in two ways which may be described as element anisotropy due to material and element anisotropy due to geometry. An example of the former is the case of anisotropic ferrite or gaseous elements of any shape immersed in a magnetostatic field while an example of the latter is the case of metallic or dielectric objects of nonspherical shape. There can, therefore, in the most general case be three orders of anisotropy in a lattice—structural anisotropy at the lattice level, geometrical anisotropy at the lattice element level, and material anisotropy of the elements on a molecular level.

The particular case of ferrite spheres in a cubical array will now be considered because of the great interest shown in recent years in propagation through generalized ferrite loaded regions.

Polder<sup>6</sup> has shown that a ferromagnetic medium which is homogeneously magnetized to saturation by a magnetostatic field is characterized by a tensor permeability. That is, the radio-frequency magnetic field intensity  $\mathbf{h}$  and flux density  $\mathbf{b}$  are related by

$$\mathbf{b} = (Q) \cdot \mathbf{h}, \quad (28)$$

where  $(Q)$  is the permeability tensor and has the form:

$$(Q) = \mu_0 \begin{bmatrix} \mu & -jK & 0 \\ jK & \mu & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

when the magnetostatic field is along  $z$ -direction. The quantities

$$\begin{aligned} \mu &= \mu' - j\mu'' \\ K &= K' - jK'' \end{aligned}$$

are complex. However, many ferrites exist for which the magnetic losses are extremely small provided the orienting field within the body is kept small so that the frequency of the wave propagating in the ferrite does not approach ferromagnetic resonance frequency. In the derivation here, we shall assume lossless ferrites. Equations which give  $\mu$  and  $K$ , in terms of the applied magnetic field and the fundamental atomic constants are given by Hogan.<sup>7</sup> The permeability of free space is  $\mu_0$ .

To determine the polarizability tensor of an isolated ferrite sphere, it will be assumed that the diameter of the sphere is small compared with the wavelength of the electromagnetic field outside as well as inside the sphere, in which case the external field can be considered uniform over the volume of the sphere. Therefore, the term

<sup>6</sup> D. Polder, *Phil. Mag.* **40**, 99 (1949).

<sup>7</sup> C. L. Hogan, *Bell System Tech. J.* **31**, 1 (1952).

<sup>5</sup> N. E. J. Neugebauer, *Can. J. Phys.* **32**, 1 (1954).

( $dD/dt$ ) is dropped from Maxwell's equations and the problem may be treated in a quasistatic way. Consequently, the problem is reduced to the determination of the potential  $\varphi$  satisfying the following boundary conditions

$$(1) \quad \begin{array}{l} \nabla^2 \varphi = 0 \quad \text{outside the ferrite sphere} \\ \nabla \cdot [(\mu) \cdot \nabla \varphi] = 0 \quad \text{inside the ferrite sphere} \end{array} \quad (29)$$

$$(2) \quad \varphi^+ = \varphi^- \quad \text{across the surface of the sphere} \quad (30)$$

$$(3) \quad \mu_0 \left( \frac{\partial \varphi}{\partial r} \right)^+ = \mathbf{i}_r \cdot [(\mu) \cdot (\nabla \varphi)^-], \quad (31)$$

where  $\varphi^-$  is the potential within the sphere and  $\varphi^+$ , the potential outside the sphere.  $\mathbf{i}_r$  is the unit vector along  $r$ -direction. Assume an incident magnetic field  $\mathbf{H}_0 = H_x \mathbf{a}_x + H_y \mathbf{a}_y$  where  $\mathbf{a}_x$  and  $\mathbf{a}_y$  are the unit coordinate vectors. Then at large distances from the ferrite sphere the magnetic field must reduce to  $\mathbf{H}_0$  by virtue of the fact that the dipole fields vanish at large distances. On the basis of the solution for the magnetostatic potential of a dielectric sphere in a uniform magnetic field, it can be assumed that  $\varphi^-$  and  $\varphi^+$  have the following forms:

$$\varphi^+ = -H_x x + b_1 \frac{x^2}{r^3} - H_y y + b_2 \frac{y^2}{r^3}, \quad (32)$$

and

$$\varphi^- = c_1 x + c_2 y. \quad (33)$$

The second and fourth terms on the right-hand side of Eq. (32) represent the potential of two dipoles parallel to the  $x$ - and  $y$ -axes, respectively. This is in accordance with the anisotropic behavior of ferrites by virtue of which magnetic fields are produced along both the  $x$ - and  $y$ -axes when the incident magnetic field is oriented along either  $x$ - or  $y$ -axes. It is also important to note that the assumption of uniform fields within the ferrite is valid only for elements of ellipsoidal shape. To evaluate the coefficients  $b_1$ ,  $b_2$ , and  $c_1$  and  $c_2$ , the boundary conditions (2) and (3) are imposed at different points on the sphere; namely  $P_1(r, \theta, \psi) = (r_1, \pi/2, 0)$  and  $P_2(r, \theta, \psi) = (r_1, \pi/2, \pi/2)$ , where  $r_1$  represents the radius of the sphere. Imposing boundary condition (2) results in

$$b_1 = (c_1 + H_x) r_1^3 \quad (34)$$

$$b_2 = (c_2 + H_y) r_1^3 \quad (35)$$

while boundary condition (3), expanded in spherical coordinates, becomes

$$\begin{aligned} \left( \frac{\partial \varphi}{\partial r} \right)^+ &= [\sin^2 \theta (\mu - 1) + 1] \left( \frac{\partial \varphi}{\partial r} \right)^- \\ &+ (\mu - 1) \frac{\sin 2\theta}{2r} \left( \frac{\partial \varphi}{\partial \theta} \right)^- - j \frac{K}{r} \left( \frac{\partial \varphi}{\partial \psi} \right)^- \end{aligned} \quad (36)$$

across the surface of the sphere.

Substituting the values of the derivatives of Eqs. (32) and (33) in the above at points  $P_1$  and  $P_2$  on the surface of the sphere leads to

$$-H_x - \frac{2b_1}{r_1^3} = \mu c_1 - jKc_2 \quad (37)$$

$$-H_y - \frac{2b_2}{r_1^3} = \mu c_2 + jKc_1. \quad (38)$$

The solution of simultaneous Eqs. (34), (35), (37), and (38) gives

$$\begin{aligned} m_x &= 4\pi\mu_0 b_1 = 4\pi\mu_0 r_1^3 \frac{\mu^2 + \mu - 2 - K^2}{(\mu + 2)^2 - K^2} H_x \\ &\quad - 4\pi\mu_0 r_1^3 \frac{3jK}{(\mu + 2)^2 - K^2} H_y \\ m_y &= 4\pi\mu_0 b_2 = 4\pi\mu_0 r_1^3 \frac{3jK}{(\mu + 2)^2 - K^2} H_x \\ &\quad + 4\pi\mu_0 r_1^3 \frac{\mu^2 + \mu - 2 - K^2}{(\mu + 2)^2 - K^2} H_y. \end{aligned} \quad (39)$$

These expressions agree with those obtained by Polder<sup>6</sup> for the magnetic moment of an ellipsoid when the demagnetizing factors are replaced by  $(4\pi/3)$ , which is the value of the demagnetizing factor for a spherically shaped dielectric object. By virtue of Eq. (28),  $b_z = \mu_0 b_z$ , and therefore the  $z$ -component of the magnetic moment vector,  $m_z = 0$ . It may be noted that each component of the incident magnetic field induces two perpendicular dipole fields in time quadrature.

From Eq. (39), the magnetic polarizability tensor can be written as:

$$(\alpha_m) = \begin{bmatrix} u & -jv & 0 \\ jv & u & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad (40)$$

where

$$\begin{aligned} u &= 4\pi\mu_0 r_1^3 \frac{\mu^2 + \mu - 2 - K^2}{(\mu + 2)^2 - K^2} \\ v &= 4\pi\mu_0 r_1^3 \frac{3jK}{(\mu + 2)^2 - K^2}. \end{aligned} \quad (41)$$

The results obtained previously for the dielectric tensor of an array of anisotropic elements apply identically for the computation of the permeability tensor of such an array.<sup>†</sup> Therefore the result of Eq. (27) can be used in this application. For a cubical array the structural anisotropy tensor ( $T$ ) reduces to a scalar to be denoted by  $A$ , [see Eq. (14)] where  $T_{xx} = T_{yy} = T_{zz} = A$ . Substituting the values of ( $T$ ) and ( $\alpha$ ) in

<sup>†</sup>  $\mathbf{M}$  is a  $\mathbf{B}$ -like quantity rather than  $\mathbf{H}$ -like, where  $\mathbf{M}$  is defined by  $\mathbf{B} = (\mu_0) \cdot \mathbf{H} + \mathbf{M}$  corresponding to Eq. (3).

Eq. (27) gives

$$(k_m) = \begin{bmatrix} u' & -jv' & 0 \\ jv' & u' & 0 \\ 0 & 0 & 1 \end{bmatrix} \quad (42)$$

where

$$u' = 1 + \frac{N}{\mu_0} \frac{u + A(v^2 - u^2)}{(Au - 1)^2 - A^2v^2} \quad (43)$$

$$v' = \frac{N}{\mu_0} \frac{v}{(Au - 1)^2 - A^2v^2}$$

$(k_m)$  is the relative permeability tensor of the lattice medium,  $A$  is a constant equal to  $(N/3\mu_0)$  [see Eq. (23)], and  $u$  and  $v$  are given by Eq. (41). It is important to note that the permeability tensor  $\mu_0(k_m)$  for the lattice medium has the same form as  $(Q)$ , the permeability tensor of the infinite ferrite medium. Consequently, the quantitative description of the behavior of all phenomenon involving an infinite ferrite medium can by simple substitution be applied to the lattice medium. As an example, the phase velocities of the two counter-rotating circularly polarized plane waves, into which a plane wave traveling in the  $z$ -direction in the lattice medium can be resolved, are expressed as:

$$\beta_+ = \frac{\omega}{c'} (k_{ef}' (\mu' - v'))^{\frac{1}{2}}, \quad \left( c' = \frac{1}{(\mu_0 \epsilon_0)^{\frac{1}{2}}} \right) \quad (44)$$

$$\beta_- = \frac{\omega}{c'} (k_{ef}' (\mu' + v'))^{\frac{1}{2}}$$

$\omega$  = the angular frequency of the electromagnetic field.

The  $\pm$  signs refer to positive and negative circularly polarized waves. A positive circularly polarized wave is one which rotates in the direction of the positive current producing the dc magnetic field.  $k_{ef}'$  represents the dielectric constant of the lattice medium, and is a scalar, because the lattice is cubical and the permittivity of the infinite ferrite medium is a scalar (see Sec. I). The value of  $k_{ef}'$  is given by

$$k_{ef}' = 1 + \frac{N\alpha_e/\epsilon_0}{1 - (N\alpha_e/3\epsilon_0)}, \quad (45)$$

where  $\alpha_e$  is the electric polarizability of a ferrite sphere and is given by

$$\alpha_e = 4\pi\epsilon_0 r_1^3 \frac{k_{ef} - 1}{k_{ef} + 2}$$

$k_{ef}$  is the dielectric constant of the infinite ferrite medium and is moderately high; values of  $k_{ef}$  in the range from 10 to 20 are common.

Similarly the Faraday rotation per unit length,  $(\theta/l)$ , of the infinite ferrite lattice can be computed

from

$$\theta/l = (\beta_- - \beta_+)/2. \quad (46)$$

Thus, the macroscopic behavior of the ferrite lattice medium is completely describable, the main restriction being that the ferrite elements and spacings be small compared to the wavelength within and without the element.

### Example 2. Tetragonal Array of Conducting Disks

In Example 1, a cubical array of ferrite spheres was investigated and exemplified the case of an array possessing no structural anisotropy but only material anisotropy of the elements. This was, so to speak, lattice anisotropy on a truly molecular level. In this section, a tetragonal array of circular disks will be considered exemplifying structural anisotropy as well as element anisotropy due to the geometry of the latter.

The electric polarizability of a conducting disk is a tensor and is given by,<sup>†</sup>

$$(\alpha_e) = \begin{bmatrix} \alpha_{xx} = \frac{16}{3} \epsilon_0 r_1^3 & 0 & 0 \\ 0 & \alpha_{yy} = \frac{16}{3} \epsilon_0 r_1^3 & 0 \\ 0 & 0 & 0 \end{bmatrix},$$

where  $r_1$  is the radius of the disk. The magnetic polarizability is given by<sup>‡</sup>

$$(\alpha_m) = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & \alpha_{zz} = -\frac{8\mu_0 r_1^3}{3} \end{bmatrix}$$

For a tetragonal array the structural anisotropy tensor  $(T)$  reduces to

$$(T) = \begin{bmatrix} T_{xx} & 0 & 0 \\ 0 & T_{yy} & 0 \\ 0 & 0 & T_{zz} \end{bmatrix}$$

Substituting these values in Eq. (27) yields,

$$(k_e) = \begin{bmatrix} 1 + \frac{N\alpha_{xx}/\epsilon_0}{1 - \alpha_{xx} T_{xx}} & 0 & 0 \\ 0 & 1 + \frac{N\alpha_{yy}/\epsilon_0}{1 - \alpha_{yy} T_{yy}} & 0 \\ 0 & 0 & 1 \end{bmatrix} \quad (47)$$

<sup>†</sup> The tensor components of  $(\alpha_m)$  and  $(\alpha_e)$  are derived by Lord Rayleigh, *Phil. Mag.* 44, 28 (1897).

and

$$(\mathbf{k}_m) = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 + \frac{N\alpha_{mzz}/\epsilon_0}{1 - \alpha_{mzz}T_{zz}} \end{bmatrix} \quad (48)$$

Estrin<sup>8</sup> in a paper on disk arrays realized that the conventional Clausius-Mosotti relations, to account for interactions between elements, were not valid for such arrays, and did not include interaction consideration in his treatment of the effects of array anisotropy on the propagation of incident plane waves. He did, however, indicate that some valid interaction analysis was necessary to strengthen the theory. The above expressions, (47) and (48), supply this correction for general tetragonal arrays of disks, provided the packing is not so close as to invalidate the representation of elements as a set of dipoles. The results could be extended to any uniform lattice as discussed earlier.

To illustrate the method suggested in Sec. I for the computation of the components of the structural anisotropy tensor, the dielectric constant of a tetragonal array will be numerically evaluated and its dependence on the lattice spacing will be discussed. Consider a tetragonal array with spacings  $a=b$  and  $c$ , as shown in Fig. 1. From Sec. I,

$$T_{xx} = T_{yy} = T_{zz}' + T_{zz}'' = (N/3\epsilon_0) + (L_{xx}/4\pi\epsilon_0), \quad (49)$$

where

$$L_{xx}'' = \sum_{\mu'} \frac{3x^2 - r^2}{r^5}$$

and where the summation index  $\mu'$  extends over all elements within a sphere or cube of finite dimensions centered around the origin of coordinates, except the single element at the origin. Insertion of this value of  $T_{zz}$  into Eq. (47) gives

$$k_{ezz} = 1 + \frac{N\alpha_{ezz}/\epsilon_0}{1 - (N\alpha_{ezz}/\epsilon_0)[\frac{1}{3} + (L_{xx}/4\pi N)]} \quad (50)$$

For a circular metallic disk

$$\alpha_{ezz} = \alpha_{eyy} = \frac{2}{3}d^3\epsilon_0,$$

where  $d$  is the diameter of the disk. Substituting this value of  $\alpha_{ezz}$  in Eq. (50) yields

$$k_{ezz} = k_{eyy} = 1 + \frac{\frac{2}{3}(d/a)^3 a/c}{1 - \frac{2}{3}(d/a)^3 a/c [\frac{1}{3} + (L_{xx}a^2c/4\pi)]} \quad (51)$$

Curves (a) and (b) in Fig. 4 are plots of Eq. (51) against  $(c/a)$ , for values of  $(d/a)=0.833$  and  $(d/a)=0.714$ , respectively. Curves (a') and (b') represent the same curves where  $L_{xx}$  is taken to be equal to zero. Figure 5 is a plot of  $(L_{xx}/4\pi N)$  as a function of  $(c/a)$

<sup>8</sup> G. Estrin, Proc. Inst. Radio Engrs. **39**, 821 (1951).

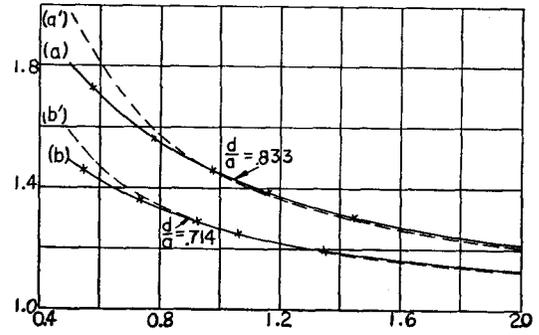


FIG. 4. Dielectric constant of the tetragonal array of disks. The crosses  $\times\times\times\times\times$  represent the experimental values.

within a cube 30 units on a side. The experimental results obtained by El-Kharadly and Jackson<sup>9</sup> are also shown in Fig. 4 and are in good agreement with the theoretical curves (a) and (b). The Clausius-Mosotti curves (a') and (b') fall below the experimental values for values of  $(c/a) < 0.9$ . This inadequacy of the classical Clausius-Mosotti relation for the disk array was first pointed out by Süsskind.<sup>10</sup> However, as demonstrated by the curves of Fig. 4, the Clausius-Mosotti relation is a good approximation for values of  $(c/a) > 0.9$  but this region is not within the range of values of interest in microwave applications. Brown and Jackson<sup>4</sup> have derived approximate relations for the dielectric constant of a tetragonal array of metallic disks by approximating the lattice by one sheet of disks when  $(c/a) > 0.6$ , and by a two-dimensional array of cylinders when  $(c/a) < 0.6$ . Their formulas are in good agreement with experiments within these ranges. However, their results are applicable only for a tetragonal array of disks and their use thereby is very limited. On the other hand, the method developed in this note is most general in that it applies for any element shape or material as well as any array geometry. The only restrictions on the method are with regard to the element size and spacing with respect to wavelength. In particular, the range of validity is determined by requirement that the element packing should not be so

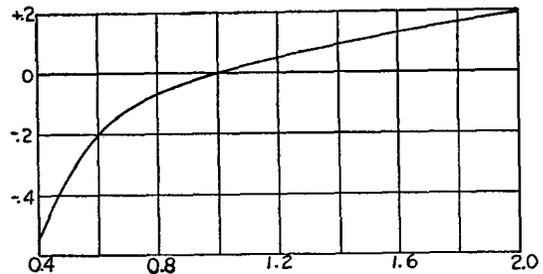


FIG. 5.  $(L_{xx}/4\pi N)$  for the tetragonal disk array.

<sup>9</sup> M. M. Z. El-Kharadly and W. Jackson, Proc. Inst. Elec. Engrs. (London) **100**, 199 (1953).

<sup>10</sup> C. J. Süsskind, "Investigation of obstacle-type artificial dielectrics," final report, Yale University.

close as to require the consideration of higher-order multipoles in the calculation of interaction effects. In the example considered, the results are valid only for values of  $(c/a) > 0.4$ .

### CONCLUSION

It has been possible to derive the constitutive dielectric parameters for uniform space arrays of generalized structural geometry composed of similarly oriented elements of completely generalized material and shape. The usual restrictions that element size and spacings be small compared to wavelength are imposed. Interactions between elements have been completely accounted for on the basis of the single additional assumption that the packing is sufficiently loose that the induced field in each element can be described by a dipolar field only. One of the results has been the

evaluation of certain conditionally convergent series representing the components of the structural anisotropy tensor ( $T$ ).

The results have been applied to two examples: A cubical array of ferrite spheres and a tetragonal array of disks. Published experimental curves for the latter show good agreement with results obtained.

The results have direct application in the design of material for the control and direction of microwaves. Work is continuing on this topic and will include a treatment for arrays of high element concentration which will require a consideration of higher-order multipoles.

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## Two Distinct Types of Short Arcs

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Short field emission arcs are of two types, those which vaporize metal predominantly from the anode by electron bombardment, and those in which the metal of the arc is supplied from the cathode largely by the melting of points by the field emission currents flowing through them. These are appropriately called "anode arcs" and "cathode arcs." A single anode arc erodes a pit in the anode and leaves a corresponding roughened area on the cathode. A single cathode arc, on the other hand, leaves on the cathode a widely dispersed array of individual pits which tend strongly to lie along scratch lines; in many cases no mark at all can be found on the anode after an arc of this type. Anode and cathode arcs differ in arc voltage, which is higher for cathode arcs than for anode arcs,

### ANODE AND CATHODE ARCS

**S**HORT field emission arcs are of two types, those in which the ions necessary for the arc are supplied chiefly by metal vaporized from the anode, and those in which they are supplied by metal from the cathode and by molecules of air. The most striking difference between these arcs can be observed in the marks which they make upon the anode and cathode surfaces. Photomicrographs of such marks are reproduced in Figs. 1 and 2.

The photographs of Fig. 1 resemble those reproduced earlier.<sup>1-3</sup> These marks were made by arcs which quite obviously melted a great deal of metal on the anode and certainly a smaller amount on the cathode. The net

and in other ways. Both types of arcs have been observed for many metals, but the data reported here are for palladium only. For clean surfaces of palladium all arcs are of the cathode type at a striking potential of 400 volts, but at 300 volts and at lower voltages anode arcs are observed also, becoming more frequent as the voltage is lowered. For surfaces activated by carbonaceous material only arcs of the cathode type occur at all striking potentials. In an anode arc, metal is transferred in both directions, as measured by radioactive tracers, but the net transfer is from anode to cathode. A cathode arc transfers metal from cathode to anode only, and the magnitude per unit of arc energy is less than the transfer in an anode arc.

transfer of metal resulting from an arc of this type is from anode to cathode. It is appropriate to call this an "anode arc."

An arc which melts metal only on the cathode, or predominantly on the cathode, can be properly called a "cathode arc." Photographs of marks made on the cathode by such arcs appear in Figs. 2(a) and 2(b). In many cases microscopic examination reveals no mark at all on the anode of one of these arcs, and in other cases it reveals a very shallow single pit such as that of Fig. 2(c). It is clear from the cathode marking that this type of discharge is made up of a great many separate arcs dispersed over a wide area of the cathode. They tend strongly to occur along scratches on the cathode surface. The excavation of the original scratches, which appears to be continuous in Fig. 2(b), can be resolved into many separate pits at higher magnification. This resolution is shown clearly in the electron

<sup>1</sup> L. H. Germer and F. E. Haworth, *Phys. Rev.* **73**, 1121 (1948).

<sup>2</sup> L. H. Germer and F. E. Haworth, *J. Appl. Phys.* **20**, 1085 (1949).

<sup>3</sup> W. S. Boyle and L. H. Germer, *J. Appl. Phys.* **26**, 571 (1955).